

Optimization of a spherical microphone array geometry for localizing acoustic sources using the generalized cross-correlation technique

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Abstract

Many workers are exposed daily to excessive noise levels. In order to reduce the noise exposure at the workstation, the main noise sources have to be detected. This task can be done with a microphone array and a source localization technique. Previous studies have shown promising results with time domain beamforming based on the generalized cross-correlation technique. The objective of this work is to propose an optimal spherical microphone array geometry dedicated to this technique. A cost function based on the symmetry of the aperture angle maps is proposed and is maximized using a Nonlinear Optimization by Mesh Adaptive Direct Search. **Numerical results show that the optimized geometry improves the noise source map by reducing the side lobe amplitude without increasing the main lobe surface.** Experimental measurements are carried out in a semi-anechoic chamber with prototyped spherical microphone arrays confirming that the optimized microphone array improves the quality of the noise source map.

Keywords: optimization, microphone array geometry, source localization, generalized cross-correlation

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1. Introduction

In Quebec, hearing loss is by far the most recorded occupational disease [1]. Many workers are exposed daily to excessive noise levels, therefore noise control solutions have to be developed to continue prevention efforts. Before proposing acoustic solutions at the workstation, the main noise sources have to be identified.

Acoustic sources can be localized using acoustic intensimetry. A camera and position tracking system can be added to monitor the position of the intensity probe. The result is a map of source positions [2]. This measurement technique works in the vicinity of the sources and is useful when the main source has been detected because it allows for localizing acoustic leaks or secondary sources. However, acoustic intensimetry is not suitable to detect and rank multiple sources in a large workplace. The other available techniques are beamforming [3] and near-field acoustic holography (NAH) [4]. These techniques have shown their capacity to accurately identify noise sources in aircraft fly-over [5], snowmobile pass-by [6] or diesel engines [7]. Beamforming and NAH have also been used to detect acoustic sources on machine tools. For example Eret and Meskell [8] have used a 25 planar microphone array to detect compressed air leaks with a beamforming technique. Zhiwei *et al.* [9] have considered NAH to detect sources on an industrial sewing machine. Bai and Chen [10] have investigated a planar microphone array to detect noise sources on a turning lathe and a milling machine. In all these cases, the microphone array was set in front of the machine.

24 In a workplace such as an industrial hall or machine shop, the number of
 25 sources is highly variable and can be distributed over all directions around
 26 the worker, therefore acoustic intensimetry and planar microphone arrays are
 27 not appropriate. Noël *et al.* [11] proposed a 15-channel spherical microphone
 28 array to detect acoustic sources in an industrial hall relying on a localization
 29 technique based on the cross-correlation functions of the microphone signals,
 30 which is similar to a time domain beamforming, and the resolution of an
 31 inverse problem. Similar works have been performed either to improve the
 32 computation time of the inverse problem or the performance of the General-
 33 ized Cross-Correlation (GCC) technique [12, 13]. Others works have shown
 34 that the GCC technique is able to identify sound source positions in a rever-
 35 berant environment or with few microphones [14, 15].

36 In the present work, the GCC technique and an open spherical array are
 37 combined to detect the acoustic source positions. The spherical array has
 38 to be small enough to be easily moved from one workstation to another and
 39 the number of microphones has to be low to ensure a fast computation time.
 40 Based on these constraints, a cost function is derived to define an optimal
 41 array geometry that will lead to accurate noise source maps.

42 **Previous studies have addressed the optimization of microphone**
 43 **array geometry. In the case of frequency-domain beamforming us-**
 44 **ing planar microphone arrays, noise source map metrics such as**
 45 **the Main Lobe Width (MLW) or the Maximum Side lobe Level**
 46 **(MSL) can be optimized [16, 17].** Planar spiral shaped arrays are a

47 common tradeoff [18, 19, 20] because they allow for reducing side lobes ampli-
 48 tude. However, these metrics depend on the source properties (position, fre-
 49 quency content, source-array distance). In the case of spherical arrays based
 50 on spherical harmonics decomposition, the microphone geometry should be
 51 chosen in order to avoid sampling errors in the computation of the spherical
 52 Fourier transform [21].

53 The spherical harmonics are not considered here and a cost function inde-
 54 pendent of the source properties is sought. Recently, a geometrical criterion
 55 has been proposed to reject side lobes and to decrease the MLW of the GCC
 56 noise source maps [13, 23]. This criterion is based on the cross product of two
 57 vectors defined by a scan point and the two positions of a given microphone
 58 pair. In this work, the angle value between these two vectors is investigated
 59 and the concept of aperture angle map is introduced. Depending on the con-
 60 sidered microphone pair, the features of the aperture angle map are different.
 61 **The spherical microphone array geometry is optimized based on a**
 62 **cost function involving the concept of aperture angle map in order**
 63 **to reduce side lobes level.**

64 The article is organized as follows. Section 2 presents the source local-
 65 ization technique based on the GCC. The aperture angle map and the cost
 66 function are introduced in Section 3 where numerical results are presented.
 67 Section 4 presents experimental data that illustrate the performance of the
 68 optimized microphone array.

69 2. Source localization technique

70 2.1. Microphone array signals

71 Consider an acoustic signal $q_q(t)$, where t is time, generated by a source
 72 located at position \mathbf{r}_q . The acoustic waves propagate in free-field conditions
 73 and reach an array of M microphones at locations \mathbf{r}_m ($m = 1, \dots, M$). The
 74 microphones record the acoustic pressures $p_m(t)$ which can be expressed as

$$p_m(t) = \alpha_{mq} q_q(t - \Delta t_{mq}) + v_m(t), \quad (1)$$

75 where α_{mq} is the geometrical attenuation due to the acoustic wave propaga-
 76 tion between the source and the microphone m and $v_m(t)$ is an uncorrelated
 77 noise due to background or measurement noise. The Time of Flight (ToF)
 78 Δt_{mq} between the source and the microphone m is given by

$$\Delta t_{mq} = \frac{1}{c_0} \|\mathbf{r}_m - \mathbf{r}_q\|_2, \quad (2)$$

79 where c_0 is the sound velocity and $\|\cdot\|_p$ is the p -norm of a vector or matrix
 80 (which are denoted by bold characters). The microphone array signal \mathbf{P} is a
 81 matrix with dimensions $[M \times N_t]$ (where N_t is the number of time samples)

$$\mathbf{P} = \begin{bmatrix} p_1(t_1) & p_1(t_2) & \dots & p_1(t_{N_t}) \\ \vdots & \vdots & \vdots & \vdots \\ p_M(t_1) & p_M(t_2) & \dots & p_M(t_{N_t}) \end{bmatrix}. \quad (3)$$

82 2.2. Generalized cross-correlation

83 Based on the microphone signals, different source localization techniques
84 can be used. In acoustics, the most standard techniques are beamforming and
85 near-field holography and they can be implemented in the time or frequency
86 domain. In this paper, time domain beamforming is chosen. The technique
87 is based on the computation of the cross-correlation function R_{mn} which is
88 typically recovered from an inverse Fourier Transform of the cross-spectrum
89 C_{mn} between microphone signals

$$R_{mn}(\tau) = \sum_{i=0}^{N_f-1} C_{mn}(i) \exp\left(j2\pi \frac{i}{N_f} \tau\right), \quad (4)$$

90 with,

$$C_{mn}(i) = P_m(i)P_n^*(i), \quad (5)$$

91 where i is the frequency index, N_f the number of frequency samples and
92 $j = \sqrt{-1}$. P_m and P_n are the fast Fourier transforms of p_m and p_n and $(\cdot)^*$
93 denotes the complex conjugate. The cross-correlation function measures the
94 similarity between two time signals. For two microphones recording the same
95 acoustic source, the maximum of the cross-correlation function estimates the
96 time delay between the two microphones. If the cross-correlation function
97 is computed for all microphone pairs, a matrix of size $[M \times M]$ is obtained
98 for each time lag τ . To minimize the computation time, the size of the
99 cross-correlation matrix $[M \times M]$ is decreased by removing the redundant
100 microphone pairs, which means the pair (m, n) is computed but the pair

101 (n, m) is not. Moreover, since the diagonal of the cross-correlation matrix
 102 provides the auto-correlation functions which correspond to a null time delay,
 103 these terms are discarded because they do not provide useful information
 104 on propagation delay. Removing these terms is equivalent to the diagonal
 105 removing of the cross-spectral matrix in the frequency domain beamforming.
 106 Finally, the cross-correlation matrix becomes a vector with size $[1 \times M_p]$,
 107 where M_p is the number of unique microphone pairs. This cross-correlation
 108 vector provides a set of time delays that depend on the microphone and
 109 source positions and the sound velocity.

110 To localize the source, a search area called scan zone has to be defined.
 111 The scan zone is a region of space which is sampled with N_s discrete points.
 112 For each scan point s and microphone pair (m, n) , a time delay can be com-
 113 puted

$$\tau_{mn}^s = \Delta t_{ms} - \Delta t_{ns}. \quad (6)$$

114 For a given microphone pair and all the scan points, the previous time delay
 115 is a vector of dimension $[1 \times N_s]$. The maximum value of this vector repre-
 116 sents the maximum time delay reachable for the considered microphone pair.
 117 Therefore, the values of the cross-correlation function evaluated at time lags
 118 above this maximum are discarded. Finally, the cross-correlation values are
 119 linearly interpolated over the scan zone. For a given microphone pair, the
 120 corresponding cross-correlation function plotted over the scan area is called
 121 Spatial Likelihood Function [13] (SLF). Each SLF corresponds to a hyperbola

122 and the location where the SLFs (each corresponding to a different micro-
 123 phone pair) intersect, is the source position. The sum of all the SLFs is called
 124 the noise source map.

125 A prefiltering operation $W(i)$ can be added to the cross-correlation func-
 126 tion Eq. 4 which leads to the GCC, denoted R'_{mn} ,

$$R'_{mn}(\tau) = \sum_{i=0}^{N_f-1} W(i) P_m(i) P_n^*(i) \exp\left(j2\pi \frac{i}{N_f} \tau\right). \quad (7)$$

127 **Among the various prefiltering techniques, the PHase Trans-**
 128 **form (PHAT) [22] is the most employed method [23, 24]. The**
 129 **PHAT operation normalizes the magnitude of the cross spectrum**
 130 **in Eq. 7 but keeps its phase unchanged.**

131 2.3. Noise source maps

132 In this section, numerical noise source maps obtained with Matlab are
 133 provided for an 18 microphones spherical array. These examples are used to
 134 introduce the cost function discussed in Section 3.

135 The microphones are located on a sphere with a 20 cm radius. Among
 136 the 18 microphones, 6 microphones are set on the equator with a 60° spac-
 137 ing. The other microphones are set at elevation angles 45° and -45° with the
 138 same azimuth angles as on the equator (see black points in Figure 1.a). This
 139 geometry is called “regular” because the microphones have the same azimuth
 140 angles. Consider a source in far-field, located at azimuth angle $\phi = 0$ and
 141 elevation angle $\theta = 0$ which generates acoustic waves in free-field conditions.

142 The sound speed is 343 m/s and the source signal is white Gaussian noise
 143 (*wgn* function in Matlab). The microphone signals are sampled at 44000 Hz
 144 and are filtered by a 2^{nd} order bandpass Butterworth filter between 100 and
 145 10000 Hz. The duration of the source signal is 1 s and the signal to noise
 146 ratio is 0 dB (white Gaussian noise is added with *awgn* function in Matlab).
 147 The scan zone extends from -180° to 180° in the azimuth direction and from
 148 -90° to 90° in the elevation direction. The grid size is (101×101) which leads
 149 to 10201 scan points. No prefiltering operation is used ($W(i) = 1$ in Eq 7).
 150 The noise source map obtained is shown in Figure 1.a. It should be noticed
 151 that the scan zone is a sphere which is unwrapped for clarity. The results
 152 at the pole are distorted due to unwrapping, for instance a horizontal long
 153 line is actually a point. The peak value of the noise source map provides the
 154 source position and corresponds to the maximum of the main lobe. In the
 155 current case, the source position is correctly located. **The MLW is usu-**
 156 **ally characterized by its width at-3 dB [19] along the observation**
 157 **directions (x and y or azimuth and elevation, here). According to**
 158 **the reference [25], the maximum of the two directions is defined as**
 159 **the MLW. Spurious lobes with a lower amplitude are also present,**
 160 **they are called side lobes. The amplitude difference between the**
 161 **main lobe peak and the closest side lobe is the MSL.**

162 If the microphone array geometry is slightly modified, the noise source
 163 map is different as shown in Figure 1.b. In this case, the 6 microphones
 164 located on the equator are shifted by -15° along the azimuth direction. This

165 small shift modifies the noise source map pattern by removing side lobes
 166 at the corners of the map and changing the side lobe pattern around the
 167 main lobe. The slices of the noise source map at $\phi = 0$ (elevation direction)
 168 and $\theta = 0$ (azimuth direction) are presented in Figure 2.a and Figure 2.b,
 169 respectively. Along the azimuth direction, the side lobe amplitude, in the
 170 vicinity of the main lobe, is decreased for the shifted configuration. The MSL
 171 difference between the regular and shifted arrays are -1.4 dB (for $\phi < 0$) and -
 172 1.5 dB (for $\phi > 0$). However, side lobes with higher amplitude appear. Along
 173 the elevation direction, the side lobe amplitude is decreased (Figure 2.b),
 174 the MSL difference are -2.3 dB (for $\theta < 0$) and -2.2 dB (for $\theta > 0$). **The**
 175 **MLW, based on the azimuth and elevation slices, remains the same**
 176 **between the two microphone configurations.**

177 In order to avoid for selecting a slicing direction as done for the
 178 MLW calculation, a covariance ellipse (with 95% confidence) is used
 179 to characterize the noise source map in this work. The covariance
 180 ellipse surrounds an area of the noise source map where the values
 181 are higher than a given threshold. The ellipse is obtained i) by se-
 182 lecting the data of the noise source map higher than the threshold
 183 and ii) by computing the covariance matrix of the data. The minor
 184 and major axes of the ellipse are given by the eigen-vectors and the
 185 radii by the eigen-values of the covariance matrix [26]. The main
 186 advantage is that the covariance ellipse takes into account the two
 187 dimensional aspect of the noise source map while the slices provide

188 an information along a unique direction. The choice of the thresh-
 189 old value allows for quantifying the main lobe or the presence of
 190 side lobes. For instance, a -3 dB threshold provides an ellipse which
 191 surrounds the main lobe, therefore it is possible to determine the
 192 main lobe surface by computing the surface ellipse. The -12 dB
 193 threshold allows for estimating the presence of side lobes. Finally,
 194 a surface reduction can be assessed by computing the relative error
 195 between two surfaces $((S_1 - S_2)/S_1)$ where S_1 and S_2 are the surfaces
 196 obtained with the regular and shifted (or optimized) array for in-
 197 stance. The covariance ellipses at -3 dB and -12 dB are shown in
 198 Figure 2.c (a zoom is provided in Figure 2.d). The main lobe sur-
 199 faces, obtained from the covariance ellipse at -3 dB are the same
 200 with the regular and shifted arrays (see Figure 2.d). In contrast,
 201 the covariance ellipse at -12 dB is reduced with the shifted array
 202 (63% of surface reduction). To conclude, a slight change in the ge-
 203 ometry can modify the noise source map. Therefore, an optimal
 204 geometry has to be found to decrease the side lobe amplitudes.

205 3. Optimization problem

206 3.1. Aperture angle map

207 The objective of the optimization problem is to find a microphone ar-
 208 ray geometry which maximizes or minimizes a cost function. In previous
 209 array geometry optimization, the MLW or the MSL were chosen as cost

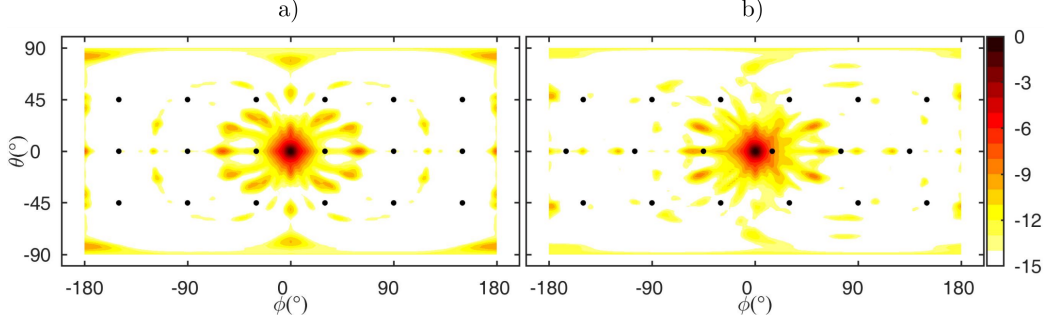


Figure 1: Noise source maps obtained with a) a regular microphone array and b) a regular microphone array where the equator microphones are shifted by -15° . The black dots are the microphone positions. The source generates white Gaussian noise and is located at $\phi = 0$ and $\theta = 0$. The colorbar is in dB. (Color online)

function [16, 17]. Reducing the MLW may affect the MSL and conversely. A multi-objective optimization can be used to tackle this problem at the expense of a more complex computation. Moreover, the MLW and MSL depend on the source position and frequency content (the higher the frequency, the smaller the MLW). Therefore, a cost function based on the resulting noise source map is not the best choice. A cost function allowing for overcoming the source influence together with an optimization procedure depending only on geometrical properties of the microphone array and scan zone are desired.

Quaegebeur *et al.* have proposed a geometrical criterion for improving noise source maps derived from GCC calculations [23]. This criterion is based solely on the scan point and microphone positions and is used in the following. Considering a scan point at position \mathbf{r}_s , it is possible to define two vectors originating from the scan point to a microphone pair position $(\mathbf{r}_m, \mathbf{r}_n)$. The angle between both vectors $\beta \in [-\pi/2, \pi/2]$, called aperture angle, can be

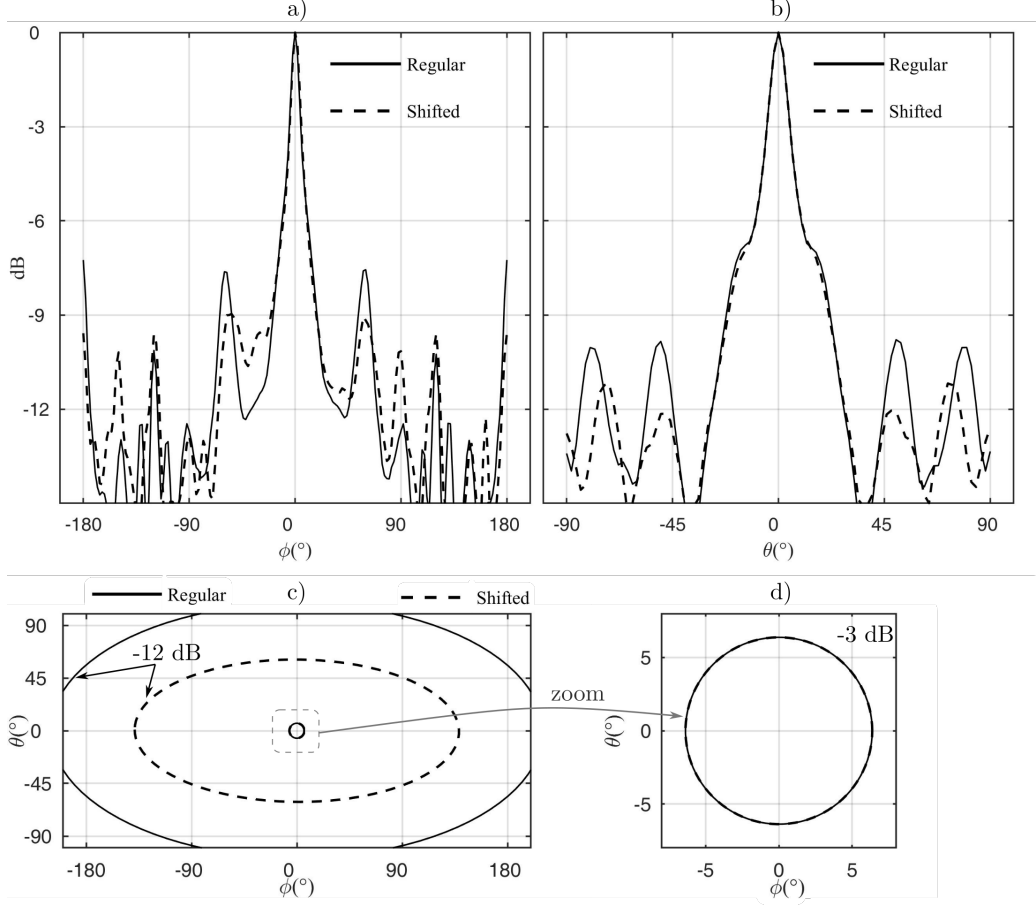


Figure 2: Slices of the noise source maps (Figure 1) along a) azimuth and b) elevation directions. c) Covariance ellipses at -3 dB and -12 dB. d) Zoom in the covariance ellipses at -3 dB.

224 computed from the normalized cross product of these vectors according to
 225 Eq. 8. Therefore, an aperture angle map $\beta^{mn}(\phi, \theta)$ can be computed for all
 226 the scan points s at angles (ϕ, θ) and a microphone pair (m, n) ,

$$\beta^{mn}(\phi, \theta) = \arcsin \left(\frac{\|\vec{\mathbf{r}}_{ms} \times \vec{\mathbf{r}}_{ns}\|_2}{\|\vec{\mathbf{r}}_{ms}\|_2 \|\vec{\mathbf{r}}_{ns}\|_2} \right). \quad (8)$$

Figure 3 shows aperture angle maps for two different microphone pairs, called pair A and pair B (microphone locations are shown by red crosses). Each aperture angle map is normalized by its peak value. In the region comprised between the two microphones, the aperture angle takes the largest values in Figure 3.a. A decrease in the aperture angle map value (see e.g. values at $\phi = \pm 80^\circ$ and $\theta = 0$ in Figure 3.a) means that both vectors $(\mathbf{r}_{ms}, \mathbf{r}_{ns})$ become collinear. Both aperture angle maps present different features. A potential link with the corresponding SLF is investigated in the following section.

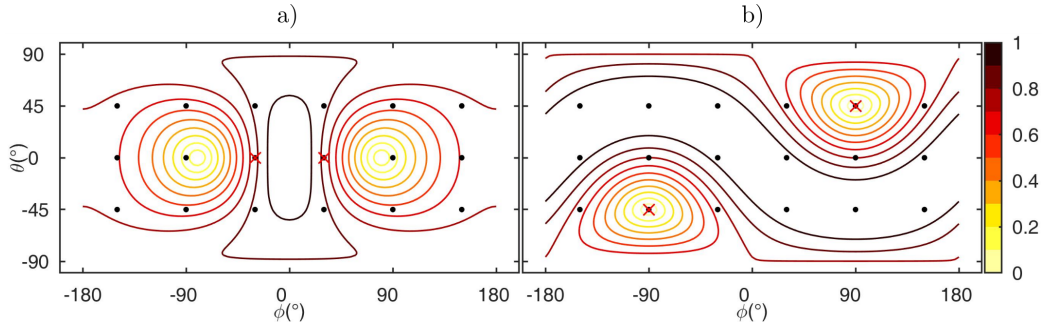


Figure 3: Aperture angle maps for the microphone a) pair A and b) pair B. The black dots are the microphone positions. The red crosses are the microphones considered. (Color online)

3.2. Spatial likelihood function

First, it is worth remembering that a SLF is obtained for each microphone pair and the noise source map is obtained by averaging the M_p SLFs. Figure 4 shows four SLFs obtained with microphone pairs A and B in the case of two different source positions. In the first case, the source is located at $\phi = 5^\circ$

241 and $\theta = 5^\circ$ where both aperture angle maps take high values (see Figure 3.a
 242 and Figure 3.b). In the second case, the source position is $\phi = -120^\circ$ and $\theta =$
 243 -25° where the aperture angle map values are smaller for both microphone
 244 pairs A and B. The source parameters are the same and identical to those of
 245 Section 2.3.

246 The SLFs consist in a main isoline (as shown in Figure 4) due to the main
 247 lobe of the cross correlation function and secondary isolines due to the side
 248 lobes of the cross correlation function. If the main isoline is large, a large
 249 main lobe may be expected or if the SLF presents secondary isolines with
 250 high amplitude, side lobes with high amplitude may be expected.

251 In the case of the microphone pair A and source position 1, the main
 252 isoline surrounds the right side of the SLF ($\theta > 0$) and a secondary isoline
 253 is apparent on the other side ($\theta < 0$). In the case of microphone pair B and
 254 source position 1, the main isoline follows a sine curve with a small width
 255 and the amplitude of the secondary isolines is low. In the case of source
 256 position 2, both microphone pairs provide a similar main isoline, however
 257 the amplitude of the secondary isolines for microphone pair A is higher.
 258 In each case, the SLFs follow the iso-contour lines of the aperture angle
 259 maps. These observations lead to the conclusion that the shape of individual
 260 SLFs is related to the angle aperture map (which is based onto the array
 261 geometry and the scan zone). Moreover, microphone pair B should provide
 262 more accurate SLFs than microphone pair A in these cases because the main
 263 isoline is narrower or the amplitude of the secondary isolines is lower. Based

on these observations, a cost function relying on the geometry of aperture
angle map is proposed.

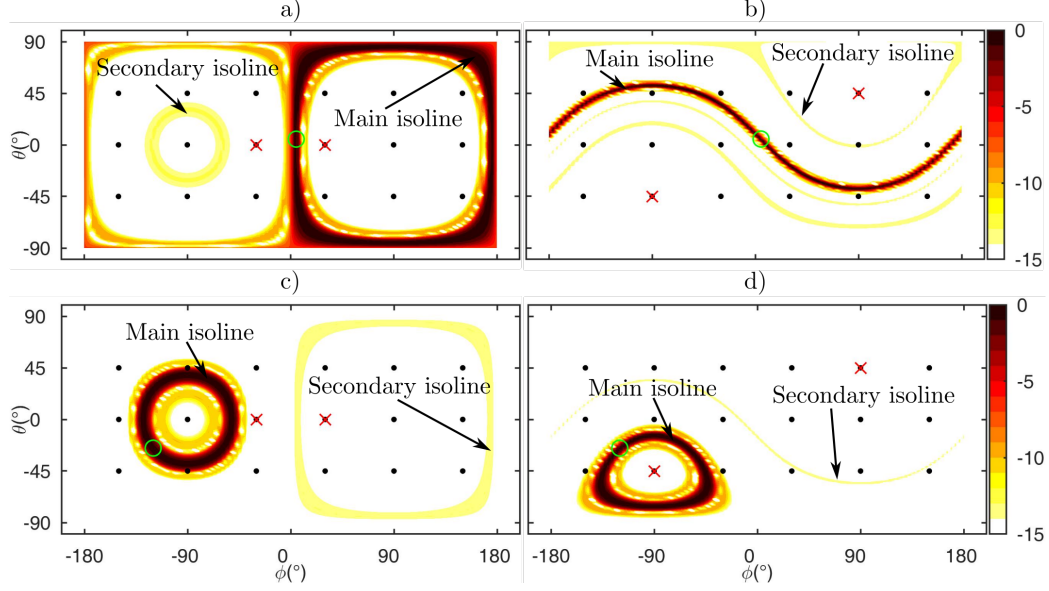


Figure 4: Spatial likelihood functions for a) pair A - source 1, b) pair B - source 1, c) pair A - source 2 and d) pair B - source 2. The black dots are the microphone positions. The red crosses are the microphones considered. The green circle is the source position. The colorbar is in dB. (Color online)

3.3. Cost function

In this section, the symmetry of the aperture angle map is investigated. With the microphone pair A, horizontal and vertical symmetries of the aperture angle map can be observed. These symmetry properties are not present in the aperture angle map obtained with the microphone pair B. To evaluate the symmetry of the aperture angle map $\beta^{mn}(\phi, \theta)$ for the microphone pair (m, n) , two variables, denoted J_h and J_v , are computed

$$J_h^{mn}(\phi) = \frac{1}{N_\theta} \sum_{i=1}^{N_\theta} |\beta^{mn}(\phi, \theta_i) - \beta^{mn}(\phi, -\theta_i)|, \quad (9)$$

273

$$J_v^{mn}(\theta) = \frac{1}{N_\phi} \sum_{i=1}^{N_\phi} |\beta^{mn}(\phi_i, \theta) - \beta^{mn}(-\phi_i, \theta)|, \quad (10)$$

274 with N_ϕ and N_θ the number of scan points along each direction. The resulting
 275 vectors size are $[1 \times N_\phi]$ and $[1 \times N_\theta]$ for $J_h^{mn}(\phi)$ and $J_v^{mn}(\theta)$ respectively. If
 276 the aperture angle map is symmetric with respect to θ (respectively ϕ), then
 277 $J_h^{mn}(\phi)$ (respectively $J_v^{mn}(\theta)$) is zero.

278 To obtain the cost function, Eq. 9 and Eq. 10 are averaged

$$J^{mn} = \frac{1}{N} \sum_{i=1}^N J_h^{mn}(\phi_i) + J_v^{mn}(\theta_i), \quad (11)$$

279 where N is the number of point to perform the average. If the number of
 280 scan points is different along azimuth and elevation directions, the horizontal
 281 or vertical variables are interpolated to obtain the same vector length. In
 282 this case, the vector length is similar for each direction and equal to $N_\phi =$
 283 $N_\theta = N = 101$ scan points. Low values of the cost function refer to aperture
 284 angle maps with high symmetry relative to θ and ϕ , whereas higher values
 285 correspond to aperture angle maps with high anti-symmetry.

286 The cost function J^{mn} calculated for the regular array of 18 microphones
 287 presented in Section 2.3 and sorted in increasing value as a function of the
 288 microphone pairs is presented in Figure 5. Microphone pair A (considered
 289 in Figure 3.a) corresponds to the lowest value of the function. In this case,

290 the aperture angle map is fully symmetrical along both directions (θ and ϕ).
 291 Microphone pair B (considered in Figure 3.b) has the highest value. In this
 292 case, the aperture angle map is highly anti-symmetrical. In the following,
 293 few microphone pairs (m, n) , i.e. few SLFs, are used to compute the noise
 294 source map.

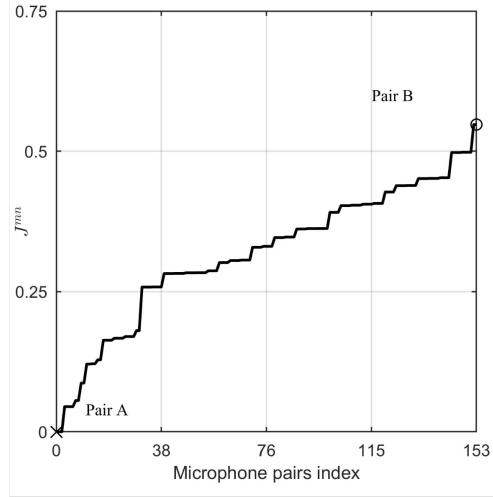


Figure 5: Cost function for the regular microphone array.

295 In order to highlight the relevance of the proposed cost function, the
 296 noise source map is now computed with only ten SLFs. The lowest, half the
 297 peak value (called average value of the cost function in the following) and
 298 highest values of the cost function are selected (see Figure 6.a-c). Two source
 299 positions are investigated, one at the origin and the other one randomly
 300 located ($\phi = 42^\circ$ and $\theta = 17^\circ$). The noise source maps obtained are compared
 301 in Figure 6.d-i.

302 When the source is at the origin, each map provides a peak value at this
 303 position (Figure 6.d-e). With the lowest values of the cost function, a vertical
 304 isoline with a high amplitude (-3 dB) is present as well as isolines surrounding
 305 the peak value (Figure 6.d). With the average values of the cost function, the
 306 side lobes are less extended and the amplitude is reduced (Figure 6.e). The
 307 side lobe amplitude is the lowest with the highest values of the cost function
 308 where only a cross is present at the source position (Figure 6.f).

309 When the source is randomly located, the noise source map obtained with
 310 the lowest values cannot localize the source position because two main lobes
 311 with the same amplitude exist (Figure 6.g). With the average values of the
 312 cost function, the source is detected but with many side lobes surrounding the
 313 source position (Figure 6.h). With the highest values, the side lobe amplitude
 314 is decreased and it is easier to localize the source position (Figure 6.i).

315 Therefore, microphone pairs that provide a high value of the cost function
 316 J^{mn} (Eq. 11) lead to more accurate noise source maps than those with a lower
 317 value. It should be noticed that the cost function proposed does not directly
 318 act on the main lobe or the side lobes of the noise source map. The cost
 319 function allows selecting properly the SLFs in order to limit the generation
 320 of a large main lobe many side lobes. It can be seen as an indirect method
 321 acting before the averaging process of the SLFs, which leads to the noise
 322 source map. To conclude, the optimization problem consists, for a given
 323 number of microphones, in finding the microphone positions which maximize
 324 the minimum value of the cost function \mathbf{J} over all possible microphone pair

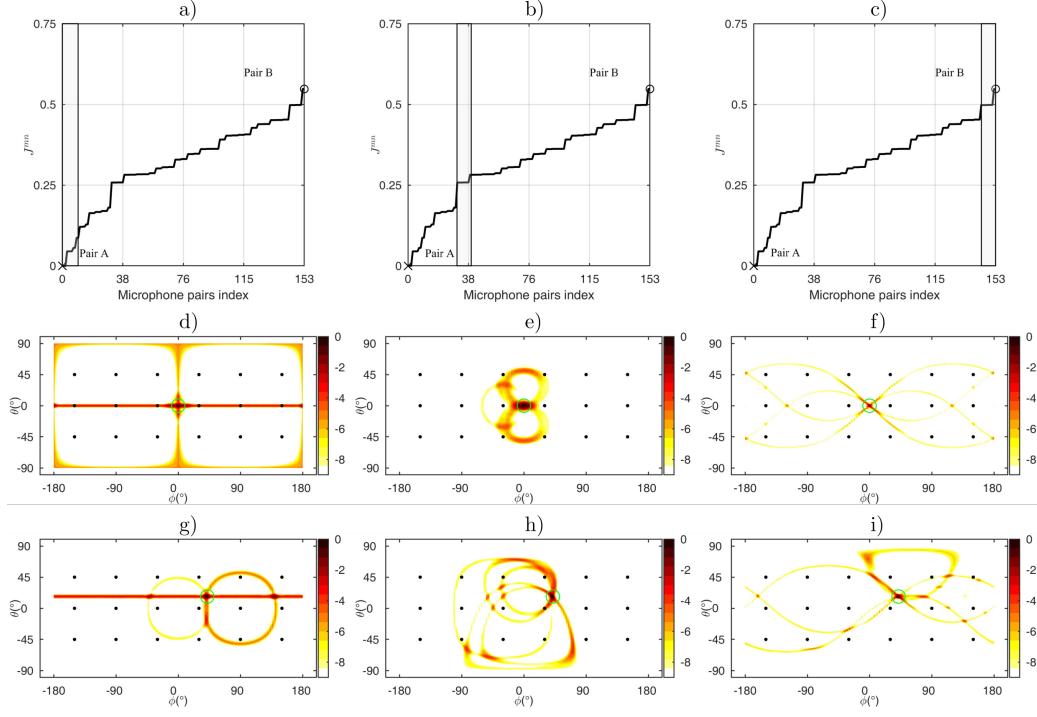


Figure 6: Cost function for the regular microphone array with a) the ten first b) the average and c) the ten last values selected. The gray patch shows the chosen microphone pairs. Noise source maps obtained with the source at origin d,e,f) and randomly located g,h,i). The black dots are the microphone positions and the green circles are the source positions. The source generates white Gaussian noise. The colorbar is in dB. (Color online)

325 positions $(\mathbf{r}_m, \mathbf{r}_n)$

$$\underset{(\mathbf{r}_m, \mathbf{r}_n)}{\text{maximize}} \min(\mathbf{J}), \quad (12)$$

326 with

$$\mathbf{J} = [J^{1,2}, \dots, J^{M-1,M}]. \quad (13)$$

3.4. Optimization solver

Now that a cost function has been defined, it is necessary to choose an optimization solver. Optimization is a dedicated topic in mathematics and this section does not aim to propose an improved solver but rather use an existing adequate and efficient solver.

Integer variables for microphone angular positions (in degrees) are considered, therefore a 1° precision is assumed. The derivatives of the cost function are not analytically accessible thus the cost function can be seen as a “black box”. A suitable solver is the Nonlinear Optimization by Mesh Adaptive Direct Search which allows for solving black box problems [28].

To avoid obtaining two microphones at the same location or all the microphones on the same side of the sphere, constraints have to be introduced. First, boundaries of the optimization space are defined to avoid microphone positions beyond $\phi = \pm 180^\circ$ and $\theta = \pm 90^\circ$. Also, each microphone is constrained to be located into a region depicted in Figure 7. Finally an inequality constraint is added. The separation between two microphones along azimuth or elevation directions has to be larger than 35° . This condition prevents two microphones to be located on the same boundary of the rectangular regions (Figure 7). No microphones are allowed beyond $\theta = \pm 70^\circ$. These regions are left empty in order to set the tripod which holds the lower surface of the spherical microphone array and a 360° camera which is placed over the top of the sphere (not used in this work).

Five distinct initial solutions are given as input to the optimization solver

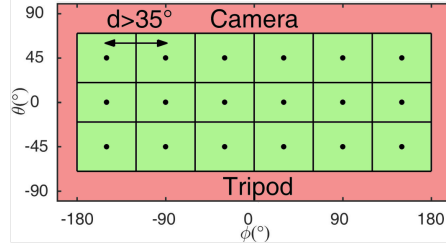


Figure 7: Regular microphone array geometry. The black lines delimit regions where individual microphones can move during the optimization. The black dots are the microphones. In red, the microphone positions unattainable and conversely in green. (Color online)

in order to compare the optimized solutions. The arbitrary selected initial microphone array geometries are

- Array 1 : Regular microphone array geometry (Figure 1.a and Figure 7)
- Array 2 : Regular microphone array geometry shifted by 10° along the azimuthal direction
- Array 3 : Regular microphone array geometry shifted by -10° along the azimuthal direction
- Array 4 : Regular microphone array geometry shifted by 20° along the elevation direction
- Array 5 : Regular microphone array geometry shifted by -20° along the elevation direction.

The cost function J^{mn} derived for the initial regular microphone array is compared with the average and min/max of the cost function derived for

363 the 5 optimized microphone arrays obtained by solving Eq. 12 (Figure 8).
 364 The five optimized arrays provide a similar trend with a low min/max de-
 365 viation. For each optimized array, the minimum value is greater than 0.2
 366 which means that the symmetrical aperture angle maps have been removed.
 367 Unfortunately, despite a similar trend for each optimized array the minimum
 368 value is different which means that the optimization solver fails to provide
 369 a global optimum. To investigate how these local solutions affect the re-
 370 sults, the noise source map and the corresponding microphone locations are
 371 compared for each optimized microphone array.

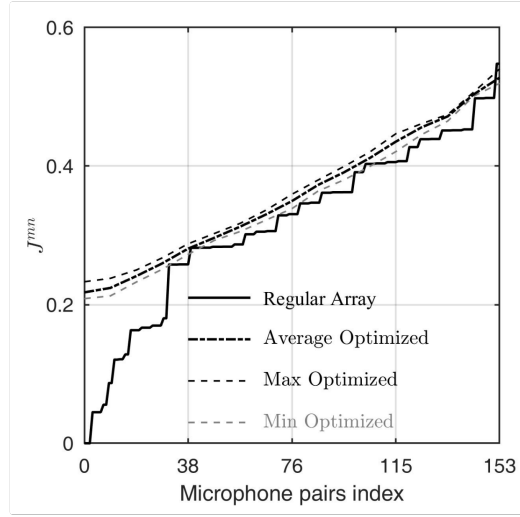


Figure 8: Cost function for the regular microphone array and for the optimized microphone arrays.

372 **Figure 9 presents the different noise source maps obtained with**
 373 **the optimized array geometries and the covariance ellipses for the**
 374 **thresholds (-3 dB and -12 dB). The side lobes are less apparent for**

375 the optimized microphone arrays than for the regular microphone
 376 array, the surface reduction is at least 85% for the -12 dB thresh-
 377 old. The side lobes at the periphery of the noise source map and
 378 surrounding the main lobe in the case of the regular array vanish
 379 for the optimized arrays. Because of the regular array symmetry,
 380 the covariance ellipse at -3 dB reduces to a circle. The optimization
 381 randomizes the microphone positions and therefore the covariance
 382 ellipses at -3 dB are no more circles with the optimized arrays
 383 which mean that the MWL is sometimes larger. However, the
 384 main lobe surface provided by the optimized arrays is at least 5%
 385 smaller. Therefore, even though different solutions are provided by
 386 the optimization solver, the side lobes are removed and the main
 387 lobe surfaces are smaller for the five optimized microphone arrays.

388 Slices of the noise source maps for optimized geometries are compared
 389 with those of the regular geometry in Figure 10. Again, only the average
 390 value and min/max deviation are plotted. In both directions, the side lobes
 391 are essentially removed. The MSL difference (obtained with the regular array
 392 and the average values) are -7.3 dB (for $\phi < 0$), -7.5 dB (for $\phi > 0$), -6.8 dB
 393 (for $\theta < 0$) and -5.8 dB (for $\theta > 0$). Therefore, solving the optimization
 394 problem Eq. 12 improves the noise source map by removing side lobes. In
 395 the following, the optimized array geometry obtained from the initial array,
 396 called Array 1, is considered.

397 To illustrate the performance of the optimized array geometry, an acoustic

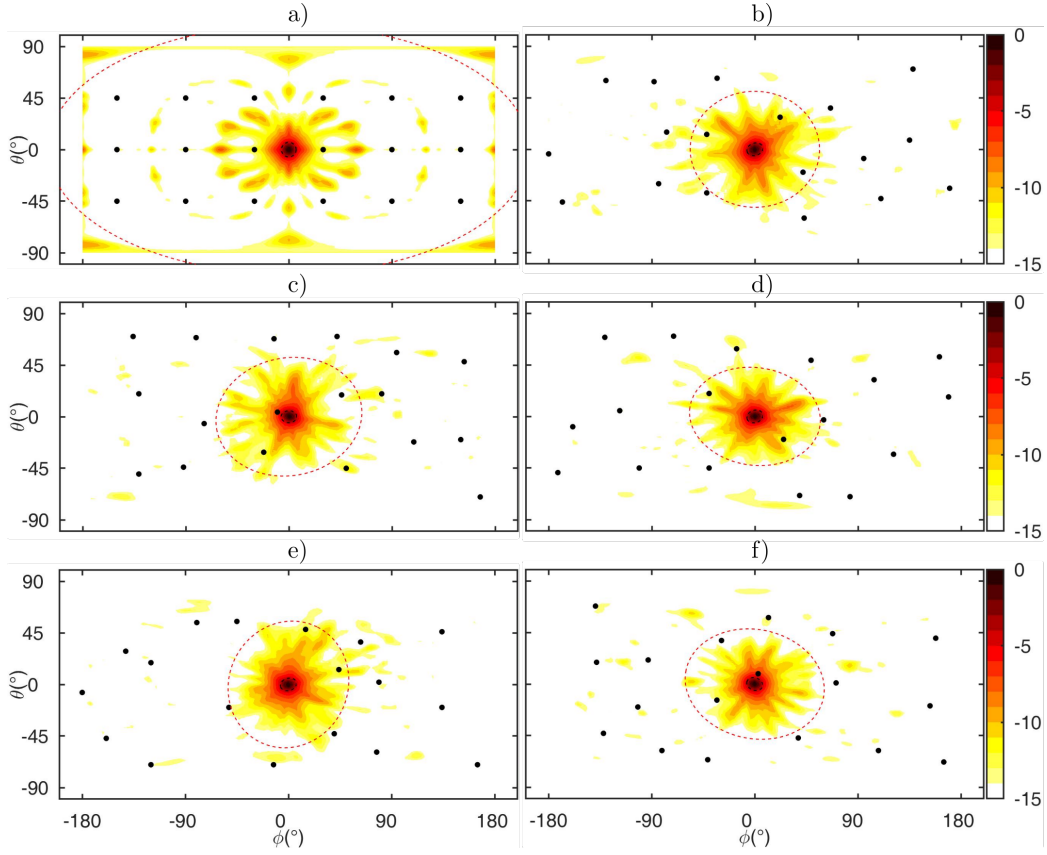


Figure 9: Noise source map obtained with the a) regular microphone array b) optimized array with initial Array 1, c) optimized array with initial Array 2, d) optimized array with initial Array 3, e) optimized array with initial Array 4 and f) optimized array with initial Array 5. The covariance ellipses at -3 dB (black) and -12 dB (red) are displayed with dashed lines. The black dots are the microphone positions. The source generates white Gaussian noise and is located at $\phi = 0$ and $\theta = 0$. The colorbar is in dB. (Color online)

398 scenario with five acoustic sources is simulated. The sources are uncorrelated
 399 white Gaussian noises and have the same amplitude. Three array geometries
 400 are compared: regular, optimized and a randomly generated array. The
 401 noise source maps obtained with the three microphone arrays are presented

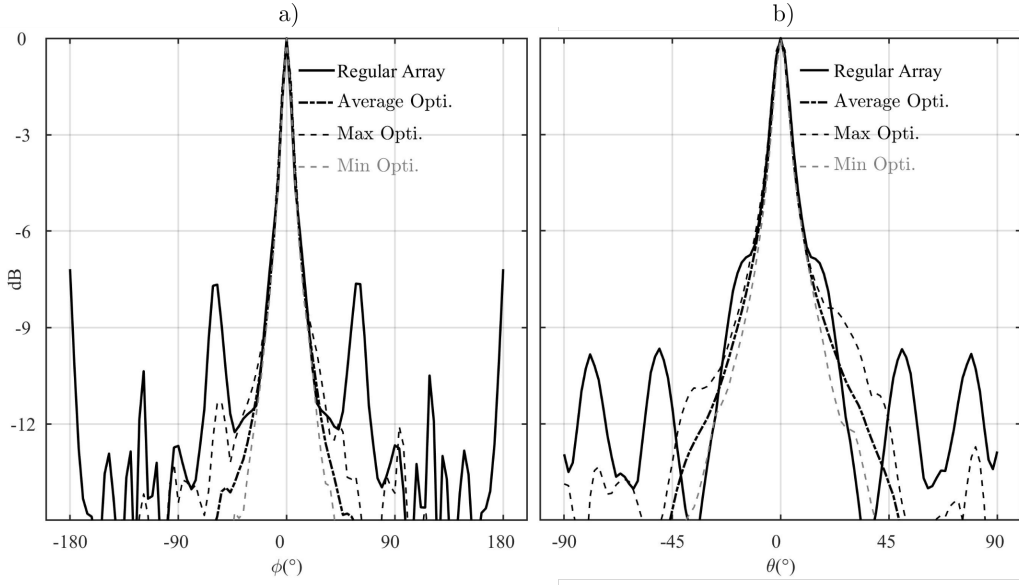


Figure 10: Slices of the noise source maps obtained with the regular array and optimized arrays along a) azimuthal and b) elevation directions.

402 in Figure 11. Each microphone array correctly detects the source positions.
 403 With the regular geometry, several side lobes are present. In a practical
 404 situation, these side lobes may be erroneously interpreted as real sources.
 405 A random microphone array improves the noise source map as some side
 406 lobes are removed. The minimum cost function value in this case is 0.08
 407 which is less than half the value obtained with the optimization solver. The
 408 optimized geometry provides the most accurate noise source map with only
 409 few side lobes.

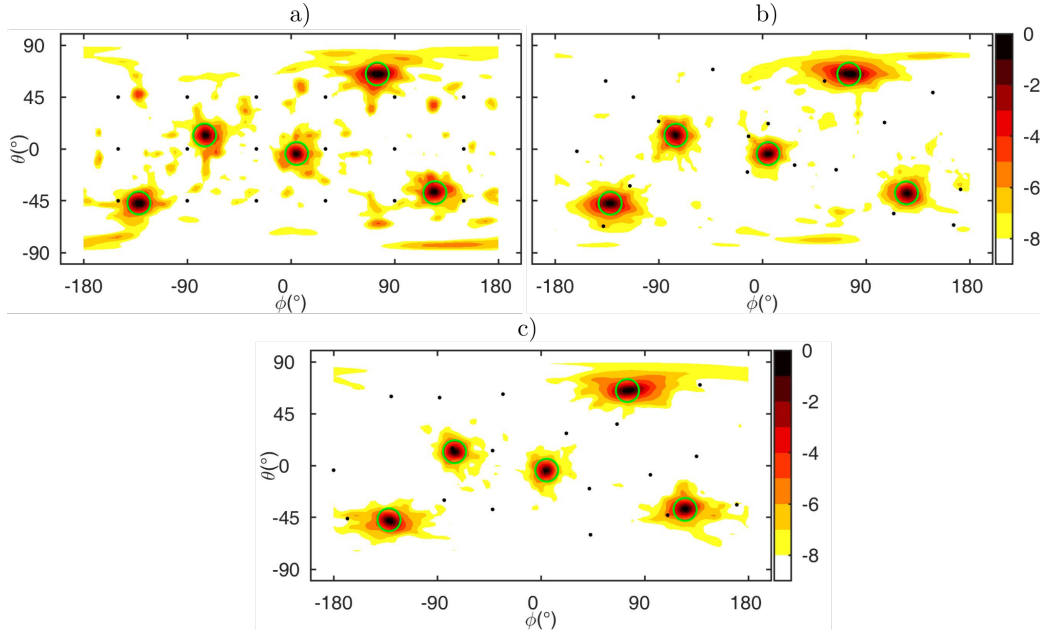


Figure 11: Noise source map obtained with the a) regular microphone array b) a random array, c) optimized array. The black dots are the microphone positions. Five sources generate uncorrelated white Gaussian noises, they are located with green circles. The colorbar is in dB. (Color online)

4. Experimental results

The goal of the experiment was to compare the noise source maps provided with the regular and the optimized array obtained with initial Array 1 (Figure 9.b). Experiments were carried out in the semi-anechoic chamber of the ICAR laboratory at École de technologie supérieure where two speakers were set in front of the spherical microphone arrays (Figure 12).

A spherical support with a 7,5 cm radius was prototyped in Acrylonitrile Butadiene Styrene (ABS) by 3D printing. Tapped holes were drilled in the spherical support to attach rods with 9,5 cm length. The microphones were

419 mounted at 3 cm from the end of the rods which leads to a total radius of
420 20 cm for the spherical microphone array (see Figure 12).

421 In order to test different microphone array geometries, 252 holes were
422 drilled into the spherical support. For $\theta = 0$, holes with 10° increment were
423 drilled along the azimuth direction. For $10^\circ \leq \theta \leq 50^\circ$ and $-10^\circ \geq \theta \geq -50^\circ$,
424 holes with 20° increment were drilled along the azimuth direction. Finally,
425 for $50^\circ \leq \theta \leq 70^\circ$ and $-50^\circ \geq \theta \geq -70^\circ$, holes with 40° increment were
426 drilled along the azimuth direction. Consequently, both geometries (regular
427 and optimized) had to be adjusted to fit onto the spherical support. However,
428 both geometries were set onto the same spherical support which has allowed
429 for simultaneous measurements (see Figure 12).

430 As the holes drilled into the spherical support were followed a regular
431 pattern a simple shift along the azimuthal direction has allowed for adjusting
432 the microphone positions with the regular array (for convenience a line of
433 three microphone has been aligned at $\phi = 0$, see Figure 13.a). However,
434 the elevation positions have slightly changed from -45° (for the numerical
435 simulations) to -40° (for experimental measurements).

436 For the optimized array, the microphone positions were set to the closest
437 positions onto the spherical support (see Figure 13.b). Consequently, the
438 microphone positions were slightly changed for both array geometries.

439 The spherical support located at 1,3 m above the floor was held by a tri-
440 pod. Brüel&Kjaer microphones type 4935 were used and the acoustic signal
441 was recorded using a Brüel&Kjaer 3038B front end and Pulse software at

the sampling rate 32768 Hz during 1 s. The source signals were uncorrelated
white noises generated with a PXI-4461 card and amplified by two active
speakers Eris Presonus E5. The source-array distance was set to 1.5 m.

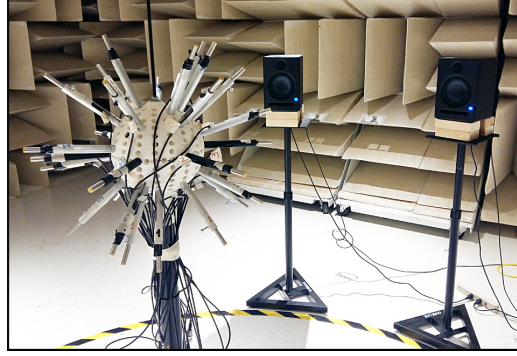


Figure 12: Picture of the spherical microphone array. (Color online)

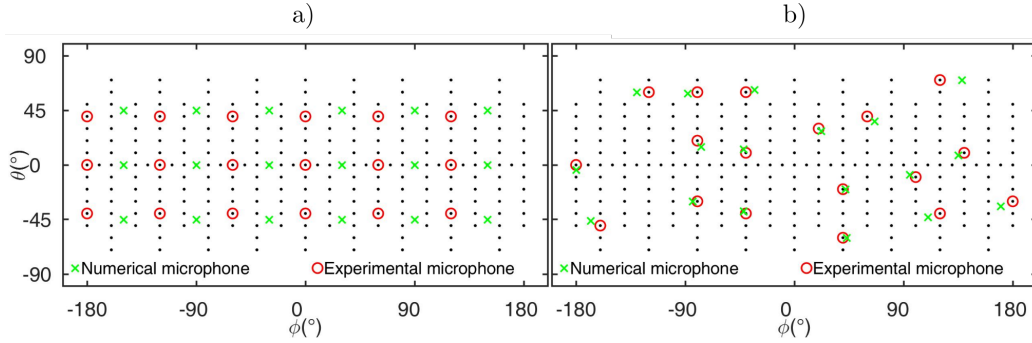


Figure 13: Numerical versus experimental microphone positions. The black dots are the positions available onto the 3D printed spherical support. (Color online)

In a first step, the source at origin was activated only to reproduce the
noise source maps obtained numerically (Figure 9.a-b), the results are pre-
sented in Figure 14. With both array geometries the source position is de-
tected. However, the regular array provides several side lobes surrounding

the main lobe whereas the optimized array removes them. Although the noise source map obtained with the experimental arrays are slightly different to numerical results of Figure 9.a-b due to slightly different microphone positions, the conclusions are the same: the optimized array provides a more accurate noise source map.

To assess the performance of the optimized array, the covariance ellipses (at -3 dB and -12 dB) provided by both array geometries are displayed in Figure 14. The optimized array allows for reducing by 66% the ellipse surface in the case of -12 dB threshold which means that some side lobes have been removed. For the main lobe surface (at -3 dB), the reduction is 14%, which means that the optimized array narrows it.

With the optimized array, a side lobe at ($\phi = 0$, $\theta = -60^\circ$) is present below the main lobe. This side lobe does not appear onto the noise source map obtained numerically and can be attributed to a floor reflection. Indeed, if the source-image reflection theory is considered, the position of the reflected source is given by $\arctan(2 \times \text{height}/\text{source-array distance}) = \arctan(2 \times 1.3/1.5) = 60^\circ$. In this case, the optimized array is able to detect the floor reflection whereas this lobe is merged with side lobes in the case of regular array.

In the last experiment, both speakers generated uncorrelated white noises with an amplitude smaller by 3 dB for the right speaker. The separation angle between the speakers was set to

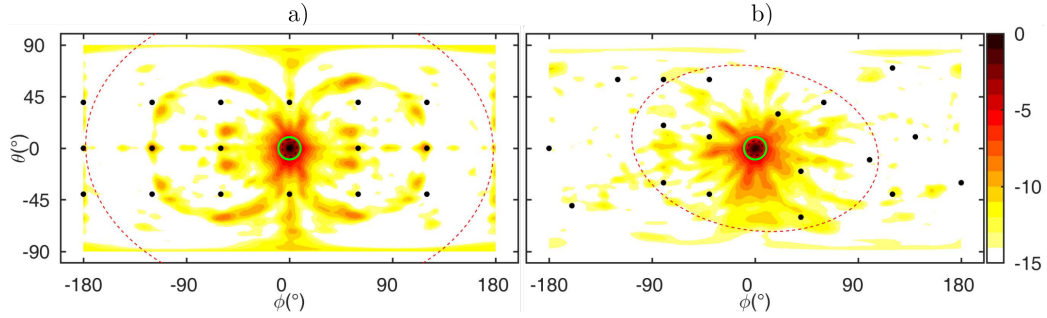


Figure 14: Noise source maps obtained with the a) regular microphone array and b) optimized array with initial Array 1. The covariance ellipses at -3 dB (black) and -12 dB (red) are displayed with dashed lines. The black dots are the microphone positions. The source was generated white noise and was located at $\phi = 0$ and $\theta = 0$. The green circle is the source position. The colorbar is in dB. (Color online)

30° which is slightly larger than the regular array resolution (Fig-
 ure 2.a-b). The noise source maps obtained with both array geome-
 tries are shown in Figure 15 together with the covariance ellipses
 (at -12 dB). The covariance ellipse at -3 dB is not computed be-
 cause two sources with different levels are considered here. Again,
 the optimized array provides a better noise source map with fewer
 side lobes. With the regular array, the side lobes amplitude is
 larger and some of them could be interpreted as sources. It is es-
 pecially the case at the opposite of the loudest source ($\phi = 150^\circ$ and
 $\theta = 0$) where side lobes intersect due to the symmetry of the array
 geometry. The surface reduction is 56% for the threshold -12 dB.
 Again, floor reflections are visible with the optimized array whereas they are
 more difficult to detect with the regular array. To highlight the performance

485 of the optimized array, the slices along the elevation direction ($\theta = 0$) are
 486 shown in Figure 16.a. With the regular array, several side lobes with am-
 487 plitudes higher than -12 dB are present whereas the optimized array limits
 488 them below -12 dB. The MSL difference are -6.8 dB (for $\phi < 0$) and -4.3 dB
 489 (for $\phi > 0$). The amplitude of the lowest source is under-estimated with
 490 both array geometries. Although the optimized array provides an amplitude
 491 closer to -3 dB the difference between both arrays in estimating the lowest
 492 source level is too small to be noticeable.

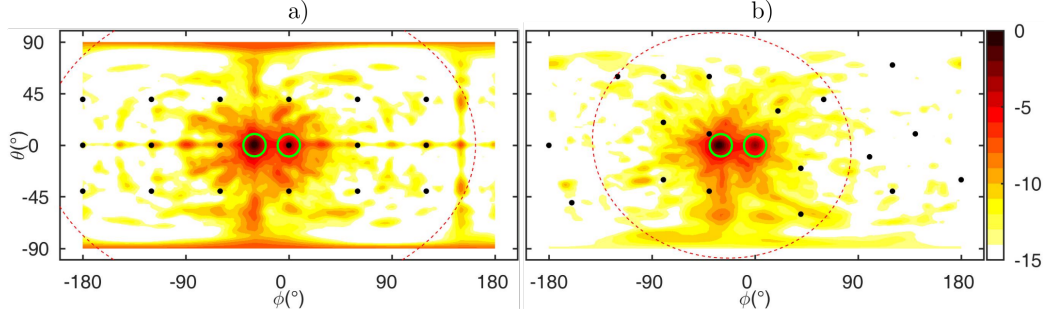


Figure 15: Noise source maps obtained with the a) regular microphone array
 and b) optimized array with initial Array 1. The covariance ellipses -12 dB
 (red) are displayed with dashed lines. The black dots are the microphone
 positions. The sources were generating uncorrelated white noises with an
 amplitude difference of 3 dB and were located at $\phi = 0$ and $\theta = 0$ and $\phi = -30^\circ$
 and $\theta = 0$. The green circles are the source position. The colorbar is in dB.
 (Color online)

493 5. Conclusion

494 Before using a microphone array-based source localization technique, the
 495 signal processing and the array geometry have to be chosen. In this work, a
 496 spherical microphone array and a source localization technique based on the

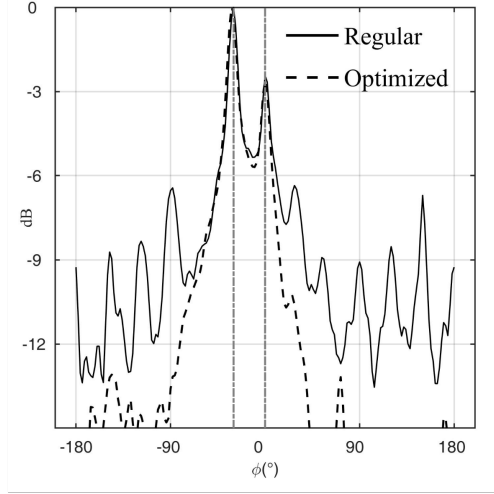


Figure 16: Slices of the noise source maps along the elevation direction. The gray dotted dashed lines are the source positions at $\phi = 0$ and $\phi = -30^\circ$.

497 generalized cross-correlation of the microphone signals have been considered.
 498 The simplest microphone configuration is the regular one where the micro-
 499 phones have the same azimuth and elevation spacing. In this case, the noise
 500 source map obtained has many side lobes which may introduce artifacts and
 501 prevent the localization of sources with weaker amplitude.

502 **In this work, the microphone positions of a spherical array were**
 503 **optimized in order to improve the source localization. A cost func-**
 504 **tion independent of the source and based on the symmetry of the**
 505 **aperture angle map was defined. The Nonlinear Optimization by**
 506 **Mesh Adaptive Direct Search solver was used to maximize it. The**
 507 **optimized microphone array geometry allows for decreasing the**
 508 **side lobes amplitudes without increasing the main lobe surface.**

509 The noise source maps obtained numerically are always more accu-
510 rate than the noise source map obtained with the regular geometry.

511

512 Finally, the regular and optimized microphone geometries have been pro-
513 totyped and experiments have been carried out in a semi-anechoic chamber.
514 The experimental and numerical results are similar and confirm that the op-
515 timized microphone array improves the quality of the noise source map by
516 removing the side lobes which allows for detecting floor reflections which is
517 not the case with the regular array.

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